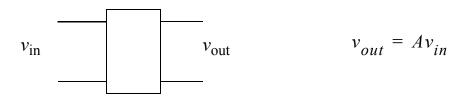
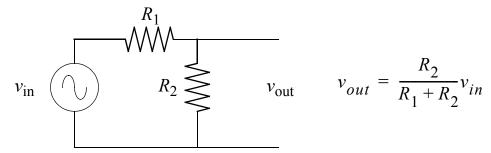
Filter Circuits -~~~ ₩

• A filter is a 4-terminal circuit element - input and output voltage, gain A.



- Input and output may both have one terminal at ground.
- A voltage divider with resistors is a filter.

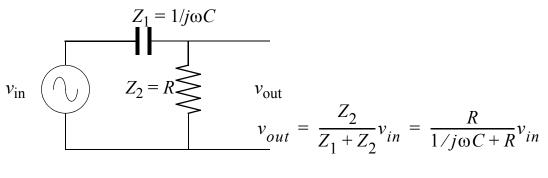


High Pass Filter

-////

- Impedances act like resistances in circuits and obey Kirchoff's laws.
- If R_1 is replaced by a capacitor the divider can be calculated with impedances.

─► <u></u>________



• This can be rewritten in terms of the complex gain *A*:

$$A = \frac{v_{out}}{v_{in}} = \frac{j\omega RC}{1+j\omega RC} = \frac{j\omega RC + \omega^2 R^2 C^2}{1+\omega^2 R^2 C^2}$$

• The terms in the denominator are comparable when $\omega_b = 1/RC$, the *breakpoint frequency*.

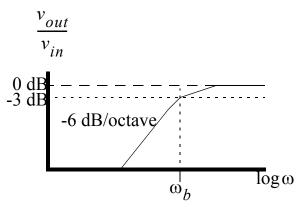
Bode Plot



• The magnitude of a complex number is $|A| = |B+jC| = \sqrt{B^2 + C^2}$ The magnitude of the gain is

$$|A| = \sqrt{\frac{\omega^2 R^2 C^2 + (\omega^2 R^2 C^2)^2}{(1 + \omega^2 R^2 C^2)^2}} = \sqrt{\frac{\omega^2 R^2 C^2}{1 + \omega^2 R^2 C^2}}$$

- For ω much less than 1/RC, A is very small. $|A| \cong \omega RC$
- For ω much greater than 1/RC, A is nearly 1.
- The high-pass filter has a breakpoint frequency $\omega_b = 1/RC$.
- A Bode plot compares log gain (dB) to log frequency.



Phase Changes ᠕᠕᠕

• The phase of the current compared to the voltage can be determined from the complex impedances: the angle in the complex plane is the phase shift

$$\tan\phi = C/B$$

• For the high pass filter:

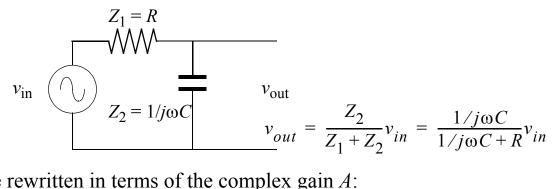
$$\sqrt{B^2 + C^2} \qquad jC = \frac{j\omega RC}{1 + \omega^2 R^2 C^2}$$
$$B = \frac{\omega^2 R^2 C^2}{1 + \omega^2 R^2 C^2}$$
$$\phi = \operatorname{atan} \frac{\omega RC}{\omega^2 R^2 C^2} = \operatorname{atan} \frac{1}{\omega RC}$$

 The phase depends on the frequency At low frequency, φ -> 90 At ω = 1/RC, φ = 45 At high frequency, φ -> 0

Low Pass Filter



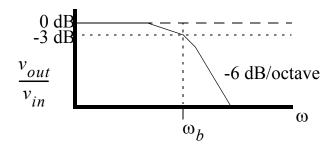
• If the capacitor replaces R_2 in a voltage divider:



• Again this can be rewritten in terms of the complex gain A:

$$A = \frac{v_{out}}{v_{in}} = \frac{1}{1 + j\omega RC} = \frac{1 - j\omega RC}{1 + \omega^2 R^2 C^2} \qquad |A| = \sqrt{\frac{1}{1 + \omega^2 R^2 C^2}}$$

- For ω much less than 1/RC, A is nearly 1.
- For ω much greater than 1/RC, A is very small.
- This is called a low-pass filter with a break frequency $\omega_b = 1/RC$. ٠
- Low pass Bode plot

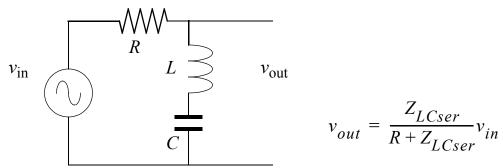


LABORATORY ELECTRONICS I





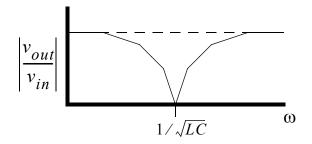
• A voltage divider with one impedance due to an inductor and capacitor in series



• The series impedance can be calculated and inserted to find the gain:

$$Z_{LCser} = 1/j\omega C + j\omega L = \frac{1 - \omega^2 LC}{j\omega C}$$
$$\frac{v_{out}}{v_{in}} = \frac{1 - \omega^2 LC}{j\omega RC + 1 - \omega^2 LC} \qquad \left|\frac{v_{out}}{v_{in}}\right| = \sqrt{\frac{\left(1 - \omega^2 LC\right)^2}{R^2 \left(1 - \omega^2 LC\right)^2 + \omega^2 L^2}}$$

• Graphically



Parallel RLC Circuit



• A voltage divider with one impedance due to an inductor and capacitor in parallel

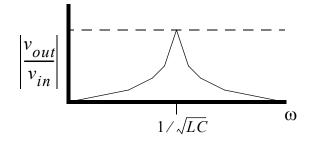
$$v_{in}$$
 V_{in} L C v_{out} $v_{out} = \frac{Z_{LCpar}}{R + Z_{LCpar}} v_{in}$

• The impedance can be calculated and inserted to find the gain:

$$Z_{LCpar} = \frac{j\omega L/j\omega C}{1/j\omega C + j\omega L} = \frac{j\omega L}{1 - \omega^2 LC}$$

$$\frac{v_{out}}{v_{in}} = \frac{j\omega L}{R(1 - \omega^2 LC) + j\omega L} \qquad \left|\frac{v_{out}}{v_{in}}\right| = \sqrt{\frac{\omega^2 L^2}{R^2(1 - \omega^2 LC)^2 + \omega^2 L^2}}$$

• Graphically

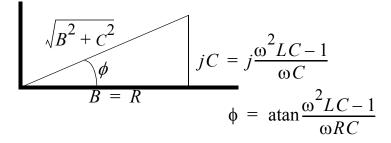


Phase Changes at Resonance

• The frequency $\omega = 1/\sqrt{LC}$ is the *resonant frequency*.

- Use $tan\phi = C/B$
- For the series RLC circuit:

$$Z_{LC} = 1/j\omega C + j\omega L = \frac{1 - \omega^2 LC}{j\omega C}$$

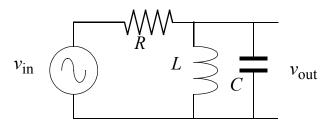


 The phase depends on the frequency At low frequency, φ -> -90 At ω² = 1/LC, φ = 0 At high frequency, φ -> +90

Quality Factor



• Consider the power gain in a parallel RLC circuit.



$$\left|\frac{v_{out}}{v_{in}}\right|^2 = \frac{\omega^2 L^2}{R^2 (1 - \omega^2 LC)^2 + \omega^2 L^2}$$

- At resonance the gain is 1.
- The frequency where the power is halved:

$$\frac{1}{2} = \frac{\omega^2 L^2}{R^2 (1 - \omega^2 LC)^2 + \omega^2 L^2}$$
$$\frac{1}{2} R^2 (1 - \omega^2 LC)^2 = \frac{1}{2} \omega^2 L^2$$
$$1 - \omega^2 LC = \frac{\omega L}{R}$$

• This has solutions at roughly:

$$\omega = \frac{1}{\sqrt{LC}} \pm \frac{R}{2L} = \omega_0 \pm \omega_0 \frac{R}{2\omega_0 L} = \omega_0 \pm \frac{\omega_0}{2Q}$$

• Q = L/R is the *quality factor* for the circuit.